

# TECH 350: DSP

Class VII: Transfer Functions, Complex Numbers,  
Prerequisites for Fourier Transform

*But first, **REVIEW?***

# Review: z-Transform

We define  $Z$  such that powers of  $Z$  correspond to the number of delays or advances in a difference equation.

One sample of delay =  $Z$ . Two samples of delay =  $Z^2$ .

One sample of advance =  $Z^{-1}$ . Two samples of advance =  $Z^{-2}$ .

Ex.:

If we're given a sequence  $x(t) = [1 \ 2 \ 3 \ 4]$ , its z-Transform,  $X(Z) = 1 + 2Z + 3Z^2 + 4Z^3$

Ex.:

If we're given a sequence  $x(t) = [1 \ 3 \ 0 \ -1 \ -5]$ , its z-Transform,  $X(Z) = 1 + 3Z - Z^3 - 5Z^4$

# Review: The Convolution Theorem

The convolution of  $f$  and  $g$  is written  $f * g$ .

An intuitive way to compute convolution by hand on 1-dimensional vectors is to think of it as a sliding window of multiplication and summation by a flipped 'kernel'. For example:

$$x_1 = (1, 5)$$

$$x_2 = (1, 2, 3)$$

$$x_1 * x_2 = (1, 7, 13, 15)$$

Convolution by hand (metaphor of sliding window):

$$\begin{array}{r} 5 \ 1 \\ \ 1 \ 2 \ 3 \\ \hline = 1 \times 1 = 1 \end{array}$$

$$\begin{array}{r} 5 \ 1 \\ 1 \ 2 \ 3 \\ \hline = 5 \times 1 + 1 \times 2 = 7 \end{array}$$

$$\begin{array}{r} \ \ 5 \ 1 \\ 1 \ 2 \ 3 \\ \hline = 5 \times 2 + 1 \times 3 = 13 \end{array}$$

$$\begin{array}{r} \ \ \ 5 \ 1 \\ 1 \ 2 \ 3 \\ \hline = 5 \times 3 = 15 \end{array}$$

# Review: The Convolution Theorem

$$x_1 = (1, 5)$$

$$x_2 = (1, 2, 3)$$

z-Transforms of these:

$$x_1(Z) = 1 + 5Z$$

$$x_2(Z) = 1 + 2Z + 3Z^2$$

Multiplying their z-Transforms together:

$$x_1(Z)x_2(Z) = 1 + 7Z + 13Z^2 + 15Z^3$$

... which is the z-Transform of the convolution of  $x_1$  and  $x_2$ !!

# Review: The Convolution Theorem

*The Convolution Theorem: convolution in the time domain is equivalent to multiplication in the z domain.*

## How does this help us?

Well, remember impulse responses of LTI systems?

***A very important application of the convolution theorem:*** any signal  $x(n)$  that is input to an LTI system, the system's output  $y(n)$  is equal to the discrete convolution of the input signal  $x(n)$  and the system's impulse response (IR)  $h(n)$ .

So, if  $x_1$  = input signal and  $x_2$  = a system's impulse response, we just figured out its output!

# LTI System: Using Input + FIR to get Output

System S is defined by the diff. equation:

$$S: y(n) = 1 x(n) + 2 x(n-1) + 3 x(n-2)$$

The FIR of S, IR(S) is: [1 2 3]

If we want the output of S, Y, given the input X = [1 2 3]:

$$IR(S) * X = [1 4 10 12 9]$$

Checking that this is equal to the output of S given X as input...

# LTI System: Using Input + FIR to get Output

System S is defined by the diff. equation:

$$S: y(n) = 1 x(n) + 2 x(n-1) + 3 x(n-2)$$

Checking that this is equal to the output of S given X as input:

$$X(1) = 1, Y(1) = 1 * 1 + 0 + 0 = 1$$

$$X(2) = 2, Y(2) = 2 * 1 + 1 * 2 + 0 = 4$$

$$X(3) = 3, Y(3) = 3 * 1 + 2 * 2 + 1 * 3 = 10$$

$$X(4) = \emptyset, Y(4) = 0 * 1 + 3 * 2 + 2 * 3 = 12$$

$$X(5) = \emptyset, Y(5) = 0 * 1 + 0 * 2 + 3 * 3 = 9$$

$$Y = [ 1 4 10 12 9 ], \text{ woo!}$$

# Transfer Functions

Give  $x(n)$  as input and  $y(n)$  as output, the *transfer function*,  $H(z)$ , may be written as:

$$H(z) = \frac{Y(z)}{X(z)}$$

Where  $X(z) = Z(x(n))$  and  $Y(z) = Z(y(n))$  (z-Transforms of  $x(n)$  and  $y(n)$ )

This follows from the fact that the impulse response convolved with the input is the output:

*$H(z)$  is equal to the z-Transform of the impulse response  $h(n)$  (neat!)*



# Transfer Functions

*Transfer Function*: provides an algebraic representation of an LTI filter in the *frequency domain*

Before we look at an example:

There are two different conventions for denoting delays and advances:

**Laplace**: uses positive powers (e.g.  $Z^2$ ) for delay and negative powers (e.g.  $Z^{-2}$ )

**Engineering**: uses negative powers (e.g.  $Z^{-2}$ ) for delay and positive powers (e.g.  $Z^2$ )

I felt that positive powers was useful for explaining the z-Transform, but we will proceed with the engineering convention moving forward

# Transfer Function: Example 1

**Difference Equation:**  $y(n) = a x(n) + b x(n-1) + c y(n-1)$  (remember a, b, and c are coefficients)

**Transfer Function:**

first, get the z-Transform of the part of the diff. equation with  $x(n)$  (for the denominator):

$$Z( a x(n) + b x(n-1) ) = 1a + bZ^{-1} = a + bZ^{-1}$$

second, get the z-Transform of the part of the diff. equation with  $y(n)$  (for the numerator):

$$Z( c y(n-1) ) = 1 - cZ^{-1} \text{ (} y(n) \text{ is on the left side, we move } c y(n-1) \text{ to the left)}$$

third, combine these:

$$H(n) = 1 - cZ^{-1} / a + bZ^{-1}$$

# Transfer Function: Example 2

**Difference Equation (biquad):**  $y(n) = a_0x(n) + a_1x(n-1) + a_2x(n-2) - b_1y(n-1) - b_2y(n-2)$

first, get the z-Transform of the part of the diff. equation with  $x(n)$  (for the denominator)

second, get the z-Transform of the part of the diff. equation with  $y(n)$  (for the numerator)

third, combine these

# Zeros + Poles (after JOS)

We may write a general form of the transfer function (with the leading coefficient in the numerator called  $g$ ) as such:

$$H(z) = g \frac{1 + \beta_1 z^{-1} + \dots + \beta_M z^{-M}}{1 + a_1 z^{-1} + \dots + a_N z^{-N}}$$

We may then factor the numerator and denominator to obtain:

$$H(z) = g \frac{(1 - q_1 z^{-1})(1 - q_2 z^{-1}) \dots (1 - q_M z^{-1})}{(1 - p_1 z^{-1})(1 - p_2 z^{-1}) \dots (1 - p_N z^{-1})}$$

If  $z$  is set to any of the numbers  $q_1, q_2, \dots, q_M$  (what are called the *zeros*), the transfer function evaluates to 0.

As  $z$  approaches any of  $p_1, p_2, \dots, p_M$  (what are called the *poles*), the transfer function approaches infinity.

# Zeros + Poles (after JOS)

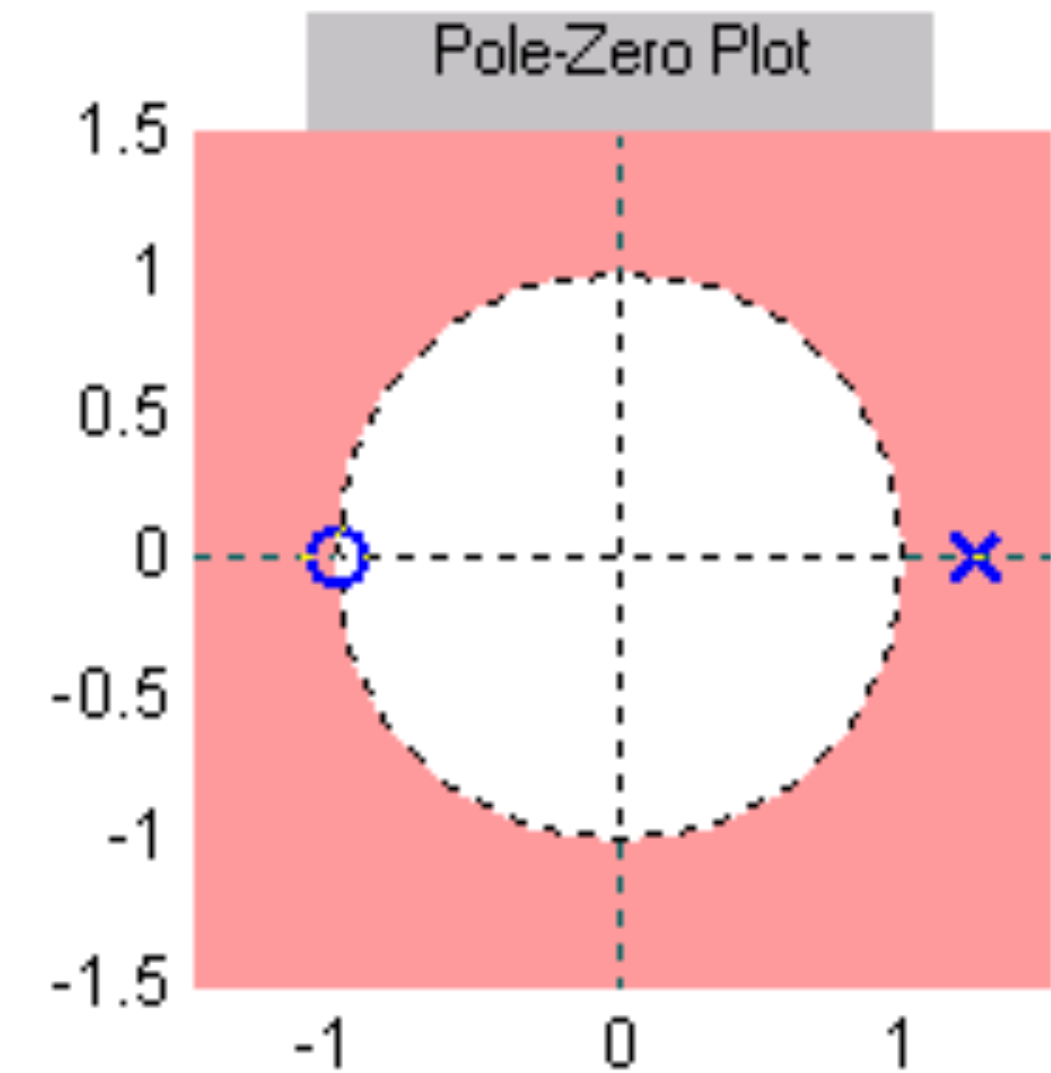
*For Computer Musicians:*

That's why, if you've screwed around with filters and coefficients, they've either suddenly gone silent (and need some coddling to get sounding again) (**zeros**) or they explode into deafening feedback (**poles**)

We can additionally use zeroes and poles to describe filters, e.g.:

*“The biquad filter is a two-pole, two-zero filter”*

$$H(z) = g \frac{1 + \beta_1 z^{-1} + \beta_2 z^{-2}}{1 + a_1 z^{-1} + a_2 z^{-2}}$$



Plot in the z-Domain that shows where the filter goes to zero (zeros, Os) and trends to infinity (poles, Xs)