



# TECH 350: DSP

Class VIII: Transfer Functions, Complex Numbers,  
Prerequisites for Fourier Transform

# Transfer Functions

Give  $x(n)$  as input and  $y(n)$  as output, the *transfer function*,  $H(z)$ , may be written as:

$$H(z) = \frac{Y(z)}{X(z)}$$

Where  $X(z) = Z(x(n))$  and  $Y(z) = Z(y(n))$  (z-Transforms of  $x(n)$  and  $y(n)$ )

This follows from the fact that the impulse response convolved with the input is the output:

*$H(z)$  is equal to the z-Transform of the impulse response  $h(n)$  (neat!)*

# Transfer Functions

*Transfer Function*: provides an algebraic representation of an LTI filter in the *frequency domain*

Before we look at an example:

There are two different conventions for denoting delays and advances:

**Laplace**: uses positive powers (e.g.  $Z^2$ ) for delay and negative powers (e.g.  $Z^{-2}$ )

**Engineering**: uses negative powers (e.g.  $Z^{-2}$ ) for delay and positive powers (e.g.  $Z^2$ )

I felt that positive powers was useful for explaining the z-Transform, but we will proceed with the engineering convention moving forward

# Transfer Function: Example 1

**Difference Equation:**  $y(n) = a x(n) + b x(n-1) + c y(n-1)$  (remember a, b, and c are coefficients)

**Transfer Function:**

first, get the z-Transform of the part of the diff. equation with  $x(n)$  (for the denominator):

$$Z( a x(n) + b x(n-1) ) = 1a + bZ^{-1} = a + bZ^{-1}$$

second, get the z-Transform of the part of the diff. equation with  $y(n)$  (for the numerator):

$$Z( c y(n-1) ) = 1 - cZ^{-1} \text{ (} y(n) \text{ is on the left side, we move } c y(n-1) \text{ to the right)}$$

third, combine these:

$$H(n) = 1 - cZ^{-1} / a + bZ^{-1}$$

# Transfer Function: Example 2

**Difference Equation (biquad):**  $y(n) = a_0x(n) + a_1x(n-1) + a_2x(n-2) - b_1y(n-1) - b_2y(n-2)$

first, get the z-Transform of the part of the diff. equation with  $x(n)$  (for the denominator)

second, get the z-Transform of the part of the diff. equation with  $y(n)$  (for the numerator)

third, combine these

# Zeros + Poles (after JOS)

We may write a general form of the transfer function (with the leading coefficient in the numerator called  $g$ ) as such:

$$H(z) = g \frac{1 + \beta_1 z^{-1} + \dots + \beta_M z^{-M}}{1 + a_1 z^{-1} + \dots + a_N z^{-N}}$$

We may then factor the numerator and denominator to obtain:

$$H(z) = g \frac{(1 - q_1 z^{-1})(1 - q_2 z^{-1}) \dots (1 - q_M z^{-1})}{(1 - p_1 z^{-1})(1 - p_2 z^{-1}) \dots (1 - p_N z^{-1})}$$

If  $z$  is set to any of the numbers  $q_1, q_2, \dots, q_M$  (what are called the *zeros*), the transfer function evaluates to 0.

As  $z$  approaches any of  $p_1, p_2, \dots, p_M$  (what are called the *poles*), the transfer function approaches infinity.

# Zeros + Poles (after JOS)

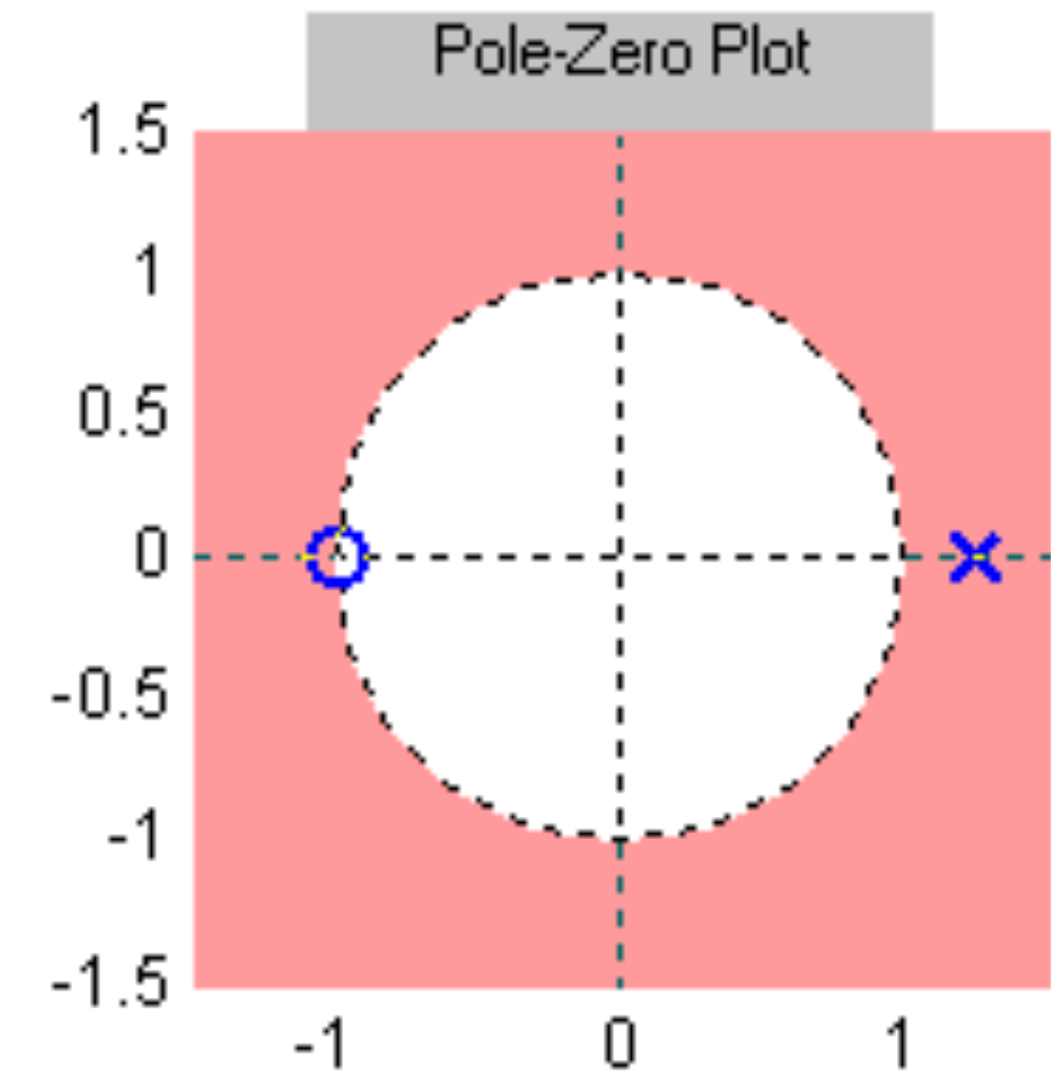
*For Computer Musicians:*

That's why, if you've screwed around with filters and coefficients, they've either suddenly gone silent (and need some coddling to get sounding again) (**zeros**) or they explode into deafening feedback (**poles**)

We can additionally use zeroes and poles to describe filters, e.g.:

*“The biquad filter is a two-pole, two-zero filter”*

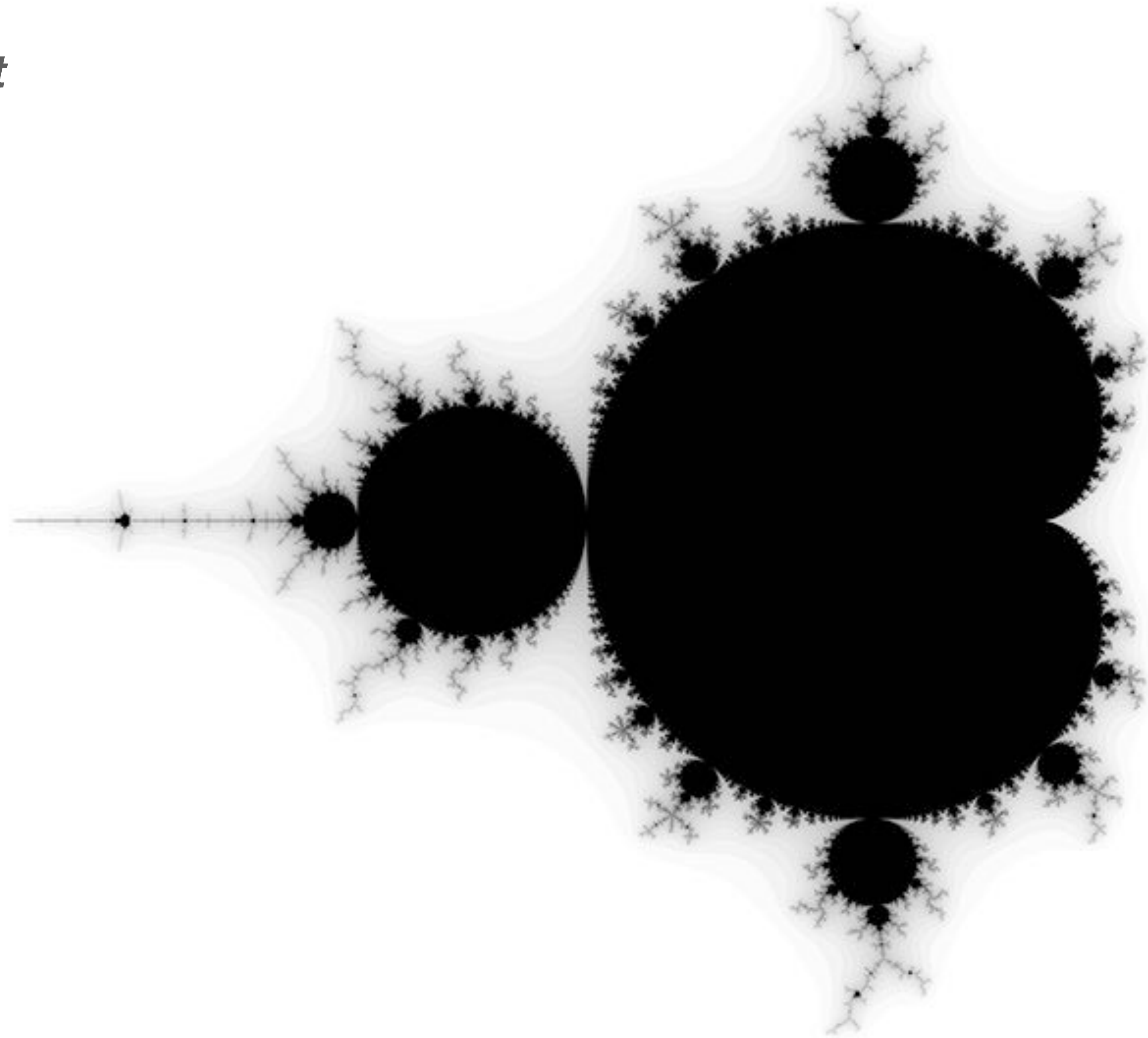
$$H(z) = g \frac{1 + \beta_1 z^{-1} + \beta_2 z^{-2}}{1 + a_1 z^{-1} + a_2 z^{-2}}$$



Plot in the z-Domain that shows where the filter goes to zero (zeros, Os) and trends to infinity (poles, Xs)



# *The Mandelbrot Set*



$$z_{n+1} = z_n^2 + z_0$$



# What is $z$ , though?

$z$  is a **complex number**, a topic that I've been avoiding, but that we necessarily need to dig into a bit before we move on to an actual definition of the Fourier Transform.

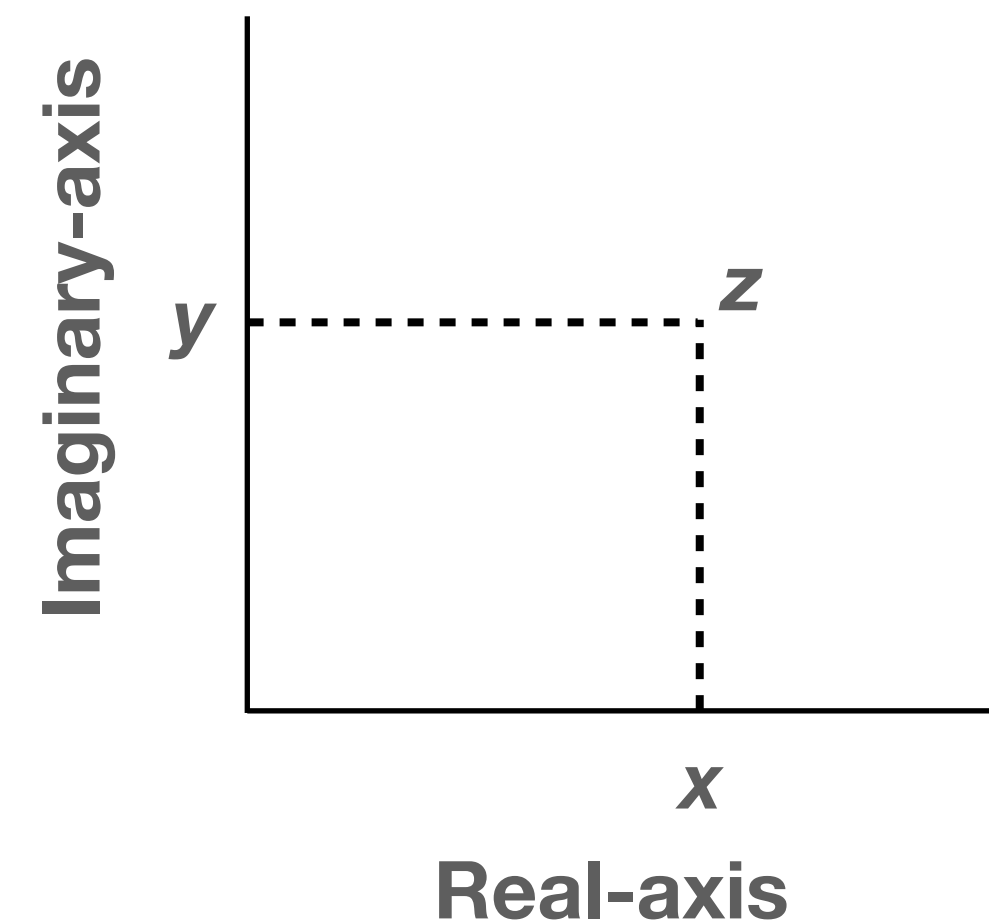
Since this class is targeted at electronic musicians, I'm going to introduce complex numbers in bite-size pieces, although some of you already eat them for breakfast.

A complex number (in its **rectangular form**) is written as:

$$z = x + yj,$$

where  $x$  is the **real** part of the number and  $y$  is the **imaginary** part  
 $j$ , which is equivalent to  $\sqrt{-1}$ , is called the **unit imaginary number**

(NOTE:  $i$  is used outside of an engineering context)



# A Taste of Complex Numbers

Converting the rectangular form ( $z = x + yj$ ) to the polar, we get the following:

$$z = r \cos(\theta) + j r \sin(\theta),$$

where  $\theta$  is an angle and  $r = |z| = \sqrt{x^2 + y^2}$

(i.e.,  $r$  is the distance from the origin of the complex number)

This is made more useful by taking advantage of Euler's identity,

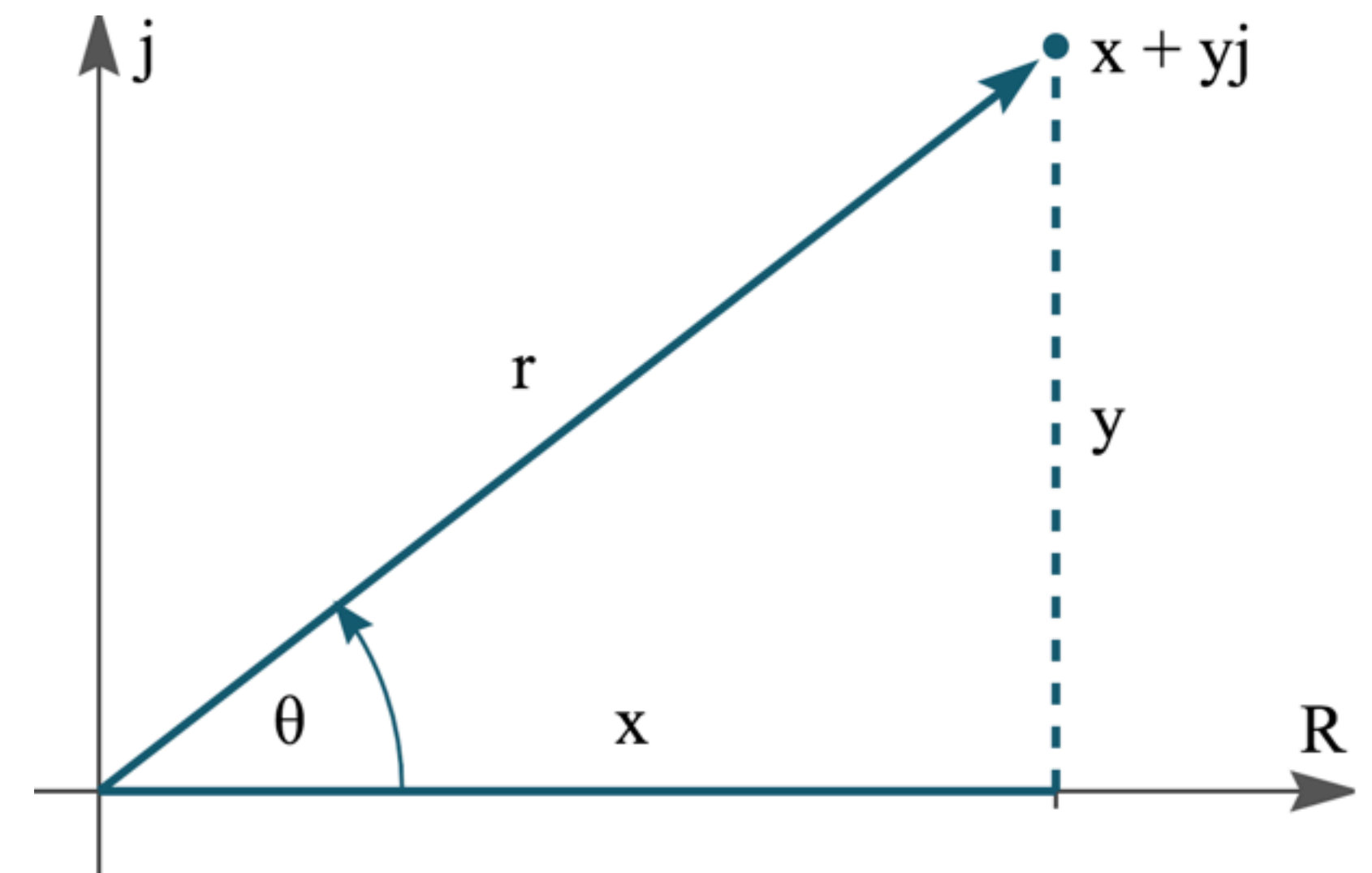
$$e^{j\theta} = \cos(\theta) + j \sin(\theta),$$

where  $e$  is Euler's number, the base of the natural logarithm,

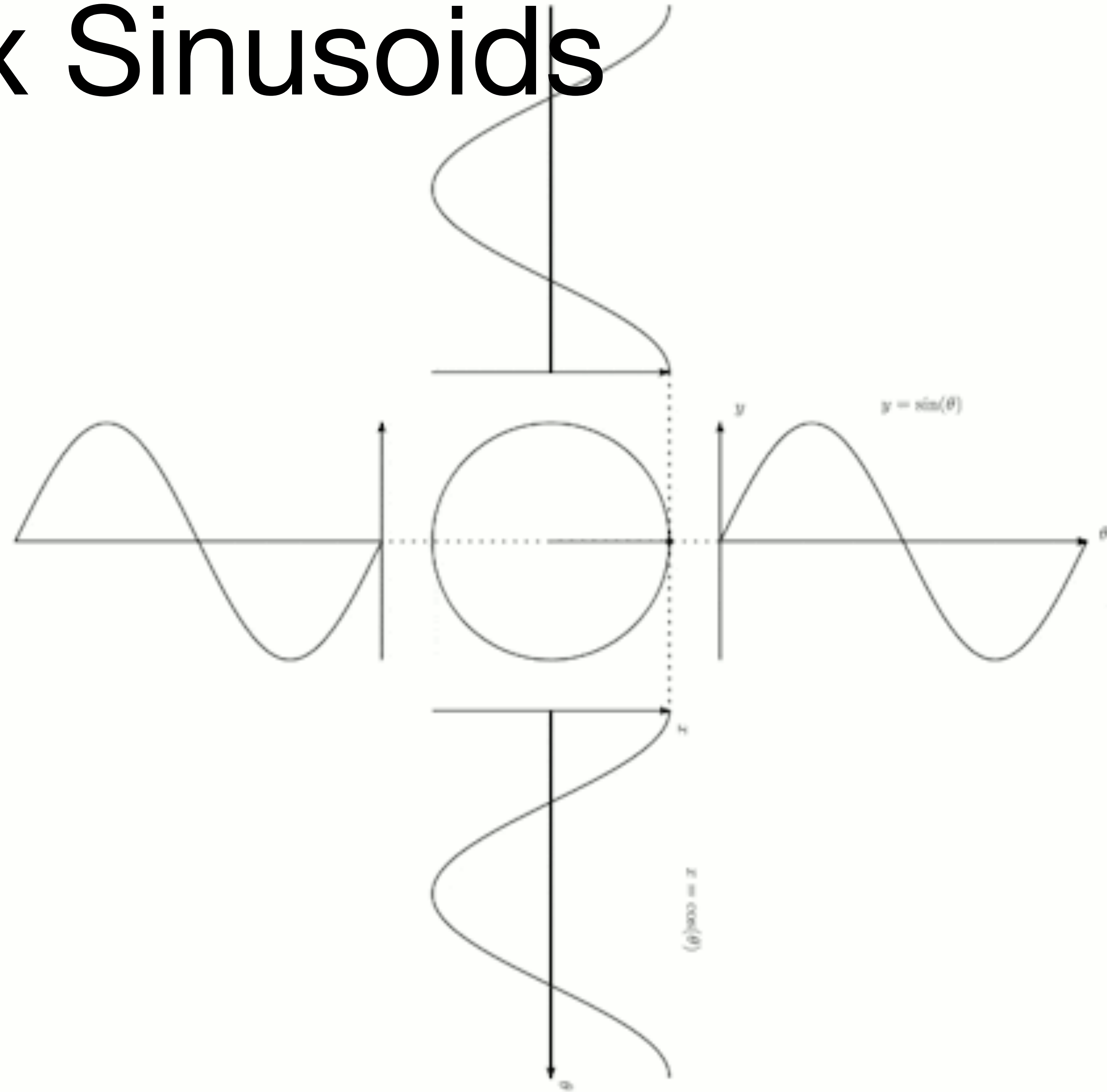
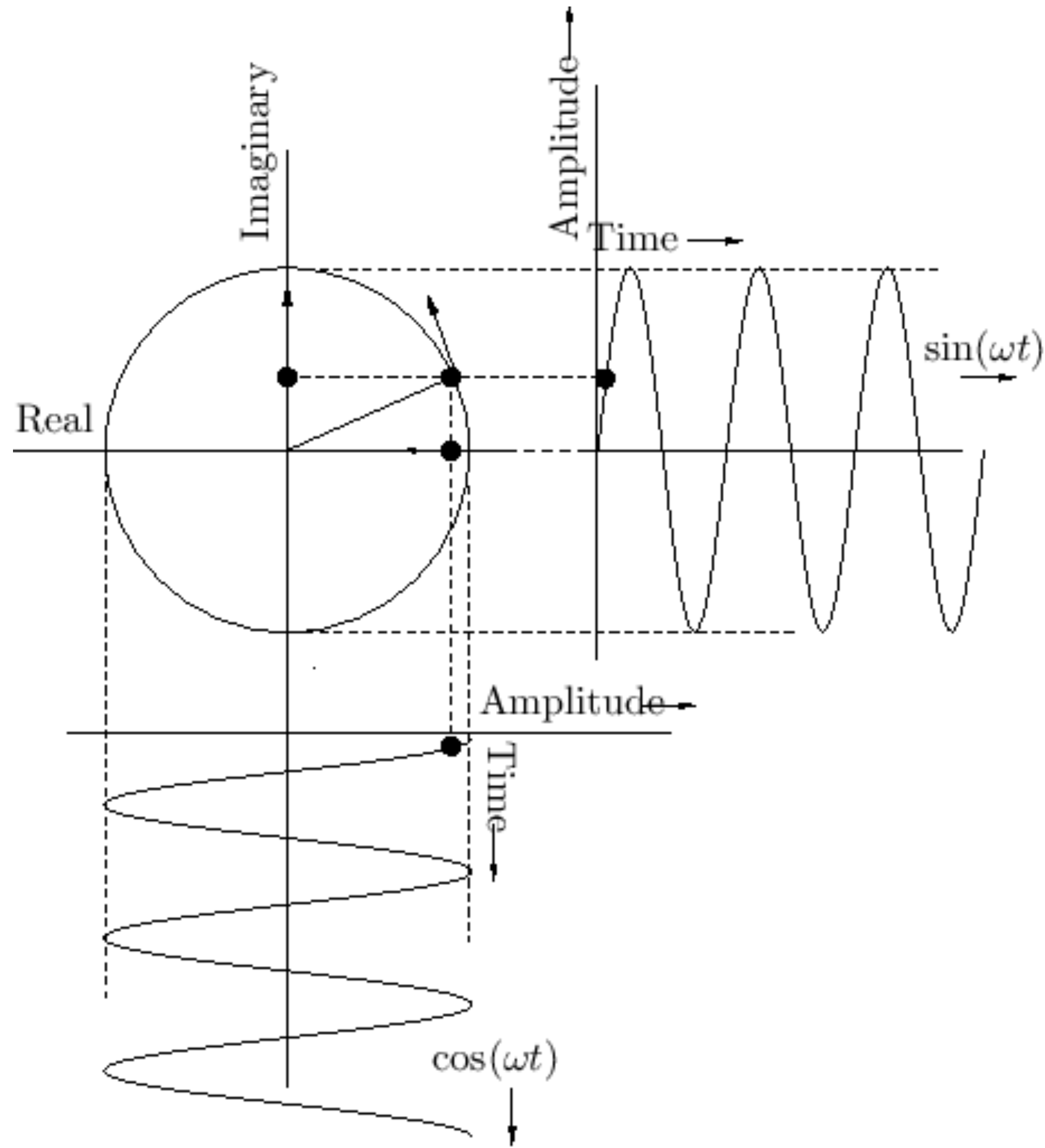
to create the form we'll use to explore the Fourier Transform:

$$z = r e^{j\theta},$$

the *polar form (AKA exponential form)* of the complex number.



# Complex Sinusoids



# Mathematical Symbolology: Summation

**Summing:** a lot of the procedures we've been doing (convolution, calculating the output of a filter) involve summing terms.

As shorthand for summation (adding together) of terms, we use the following (Big Sigma):

The diagram shows the Big Sigma notation  $\sum_{i=1}^n x_i$  with labels: "summation sign" points to the  $\Sigma$  symbol; "index of summation" points to the  $i$  below the  $\Sigma$ ; "stopping point upper limit of summation" points to the  $n$  above the  $\Sigma$ ; "starting point lower limit of summation" points to the  $1$  below the  $i$ ; and "typical element" points to the  $x_i$  term.

For example,

$$\sum_{n=1}^5 n^2 = 1^2 + 2^2 + 3^2 + 4^2 + 5^2 = 1 + 4 + 9 + 16 + 25 = 55$$



# The Fourier Transform: Take 0

## Jean Baptiste Joseph Fourier

Late 1700s researcher on periodic waves and their analysis

Studied the conductive diffusion of heat

Scientific advisor to Napoleon

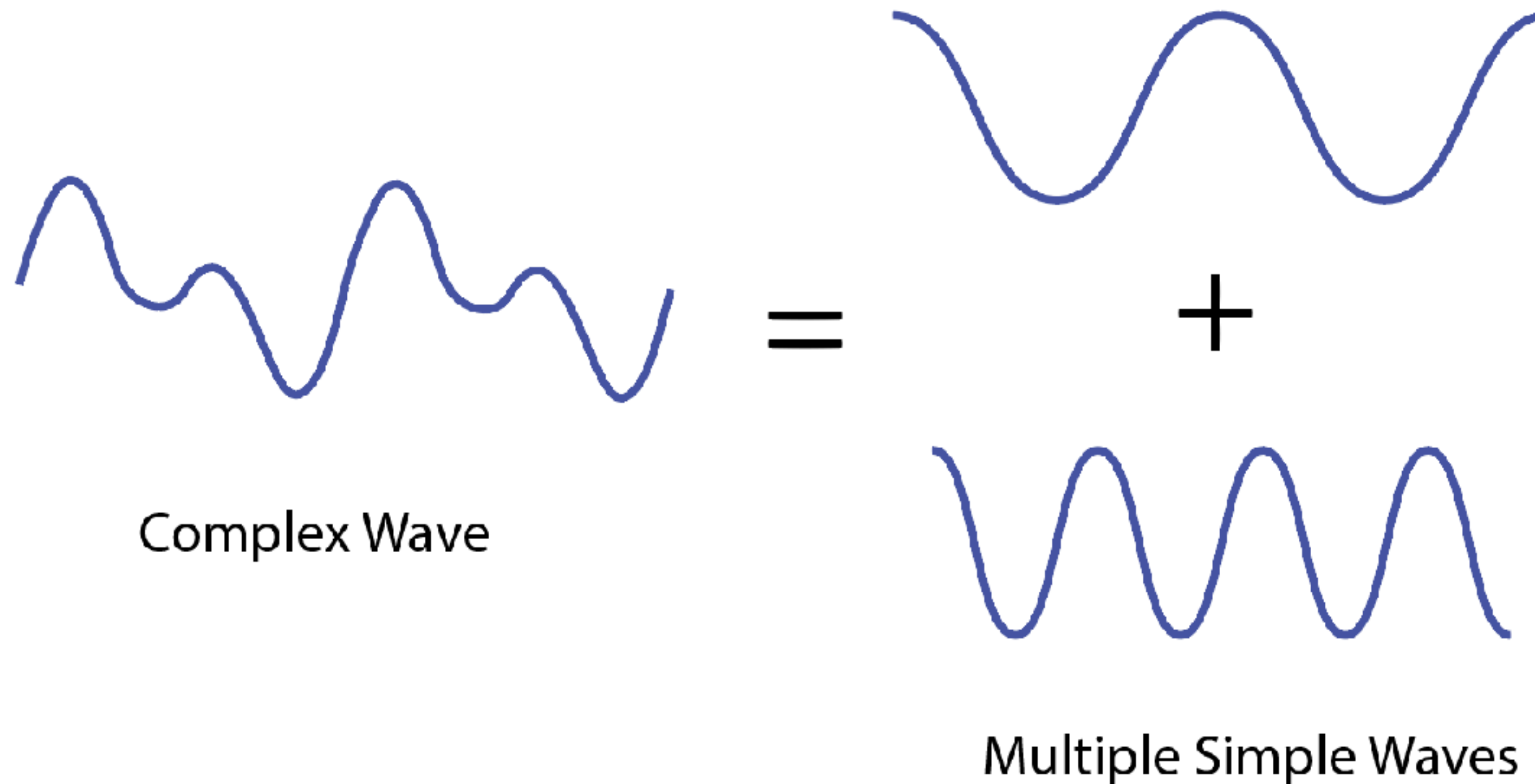




# Fourier's Theorem:

A complex periodic sound can be decomposed into a set of simple waves

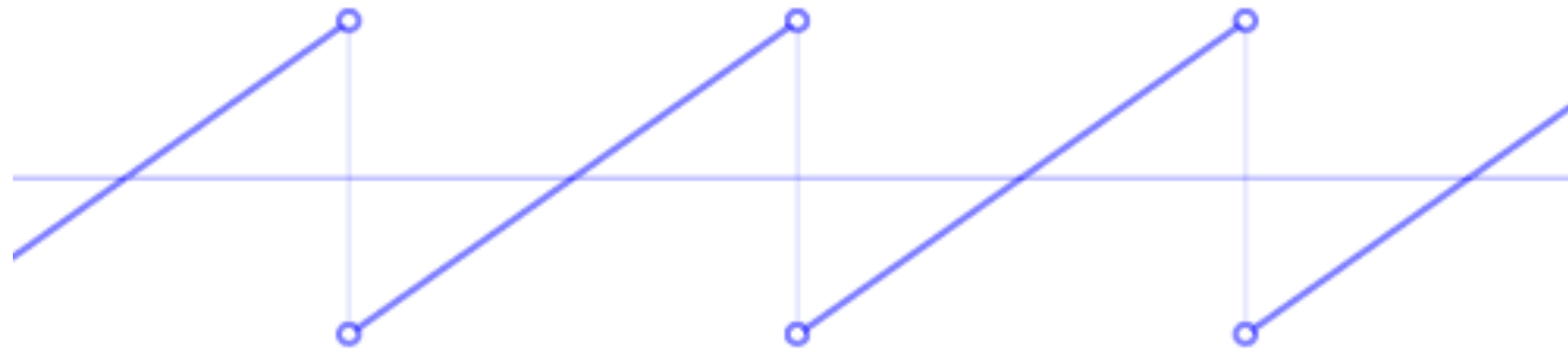
## Deconstructing a complex wave (Fourier analysis)



# Fourier's Theorem:

A complex periodic sound can be decomposed into a set of simple waves

## Summing simple waves (additive synthesis)

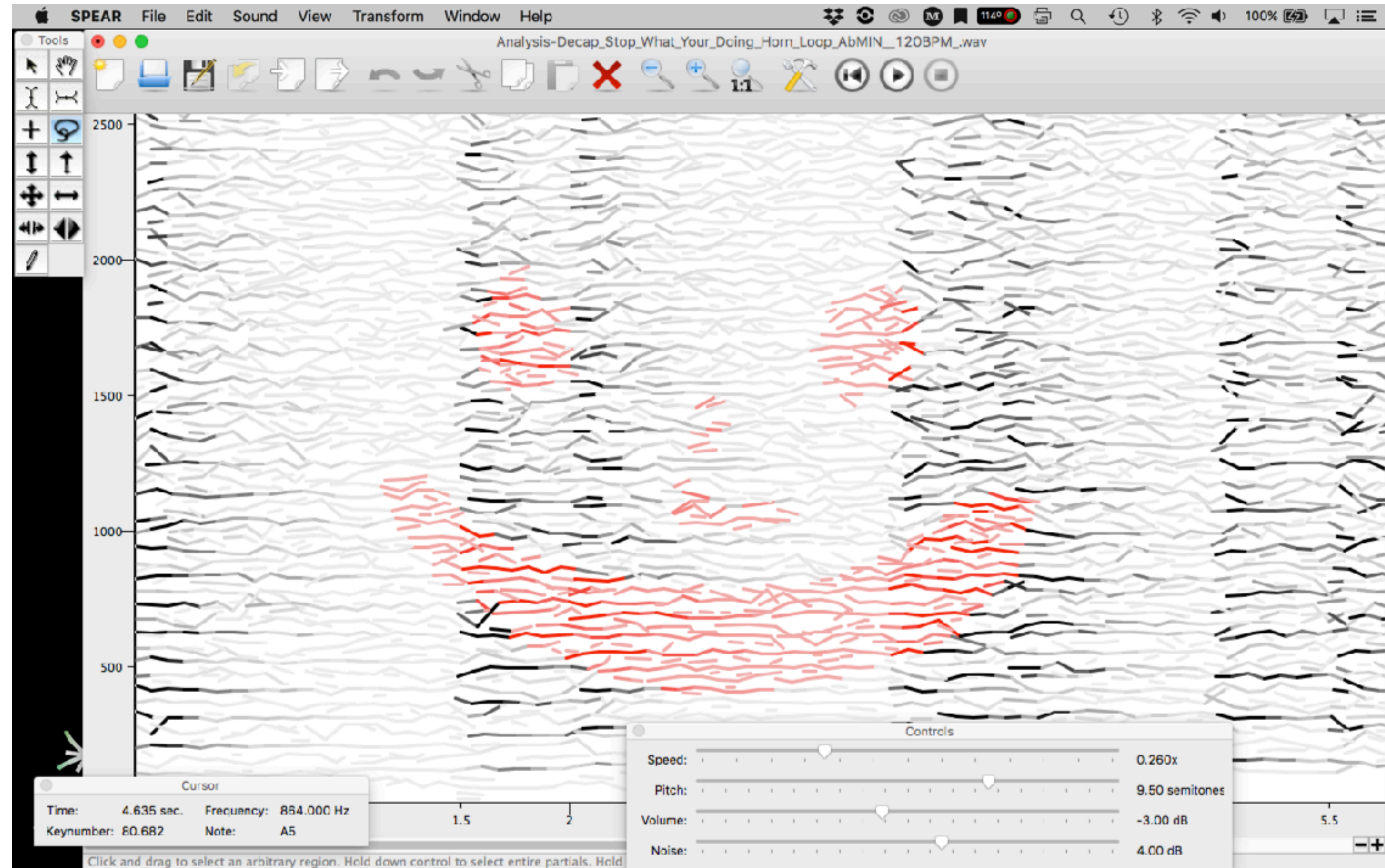


$N = 0$

Approximating a sawtooth wave using multiple simple (sine) waves



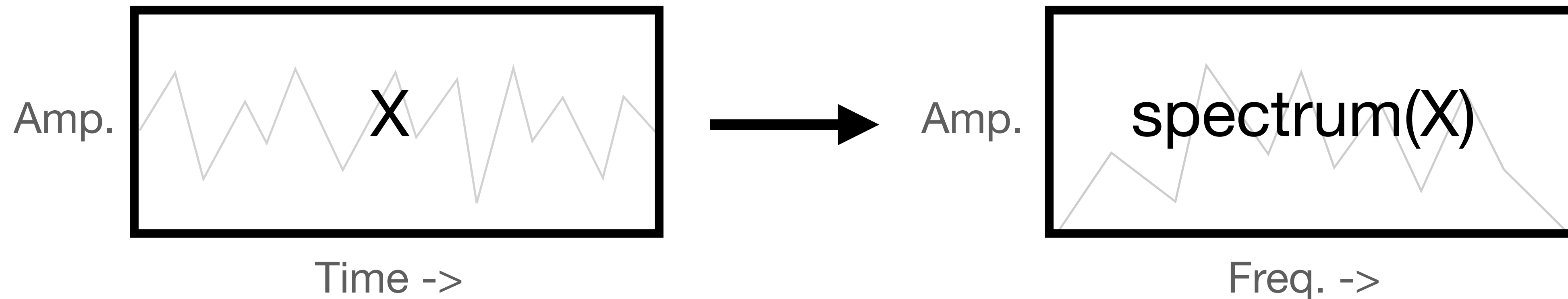
# SPEAR (Spectral Editng, Analysis, and Resyntnesis)



<http://www.klingbeil.com/spear/>

# The Fourier Transform: Take I

**Goal:** Given some signal  $X$ , we want to decompose  $X$  into its constituent frequencies, that is, we want the spectrum of  $X$ .



**The result (the Fourier Transform of  $X$ ) will give us two types of information:**

1. How much a particular frequency sinusoid is present (its magnitude) and, additionally,
2. Where in the cycle of the sinusoid it begins (its phase offset)

***Remember that we get these two different (but intimately related) types of information!***

# LTI Systems Review (Concepts we've Covered)

*System Properties*

*Signal Flow Diagrams*

*Difference Equations*

*Sine-Wave Analysis*

*Frequency Response*

*Filter Parameters*

*Filter Topologies*

*Comb + Allpass Filters*

*Digital Algorithm Reverb Basics*

*Impulse Response (FIR vs. IIR)*

*z-Transform*

*The Convolution Theorem*

*Transfer Functions*

*Zeroes + Poles*

*Complex Numbers Basics*